Weighted Parity-Check Codes

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Abstract

We introduce a new class of codes, called weighted parity-check codes, where each parity-check bit has a weight that indicates its likelihood to be one (instead of fixing each parity-check bit to be zero). It is applicable to a wide range of settings, e.g. asymmetric channels, channels with state and/or cost constraints, and the Wyner-Ziv problem, and can provably achieve the capacity. For the channel with state (Gelfand-Pinsker) setting, our code not only achieves the capacity of any channel with state, but also achieves a smaller error rate compared to the nested linear code.

Ideas and Advantages

The goal is to present a general code construction based on weighted code-book idea, but with a linear structure for efficient encoding and decoding.

- The codebook is a "fuzzy set", where each bit sequence has a weight that corresponds to the likelihood that the sequence is selected.
- By [1], weighted codebook eliminates the need of subcodebooks and gives sharper finite-blocklength and second-order error bounds.
- It applies to general(symmetric/asymmetric) channels with/without state.

Channels with State Information

Consider the channel has a state sequence $\mathbf{s} = [s_1, \dots, s_n]$, where $s_i \in \mathcal{S}$ (not necessarily binary), $s_i \stackrel{iid}{\sim} P_S$, is available noncausally to the encoder. Given \mathbf{s} , the encoder encodes message $\mathbf{m} \in \mathbb{F}_2^k$ into $\mathbf{x} \in \mathbb{F}_2^n$, which is sent through a memoryless channel $P_{Y|S,X}(y|s,x)$. The decoder receives \mathbf{y} and outputs $\hat{\mathbf{m}}$. The input may have a cost constraint $\mathbf{E}[\sum_{i=1}^n c(s_i,x_i)] \leq nD$, where $c: \mathcal{S} \times \mathbb{F}_2 \to [0,\infty)$.

Binary-Hamming Information Embedding

Consider $s_i \stackrel{iid}{\sim} \operatorname{Bern}(1/2)$ and $X \to Y$ is $\operatorname{BSC}(\beta)$. We have an expected cost/distortion constraint $\mathbf{E}[|\{i: x_i \neq s_i\}|] \leq nD$. For $0 \leq \beta \leq D \leq 1/2$, the capacity is the upper concave envelope of $H(D) - H(\beta)$.

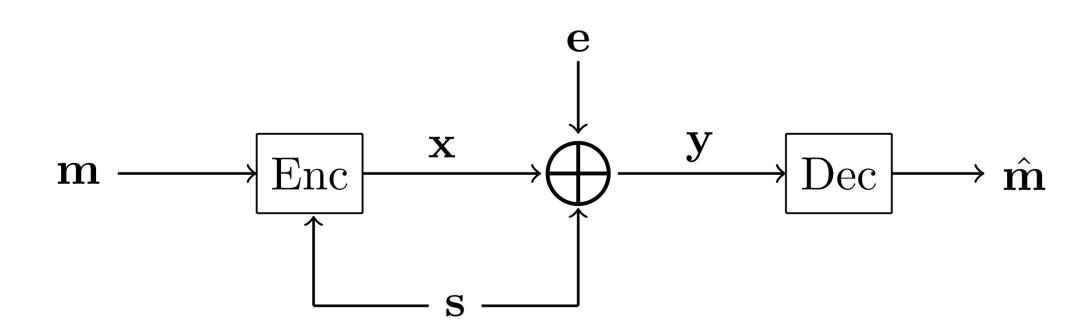


Figure 1. Block diagram of channel coding with state information. The encoder embeds \mathbf{m} into the channel input \mathbf{x} , which is with a cost constraint. $\mathbf{e} \sim \text{Ber}(\beta)$ is the channel noise.

References

- [1] Cheuk Ting Li and Venkat Anantharam. A unified framework for one-shot achievability via the poisson matching lemma. *IEEE Transactions on Information Theory*, 67(5):2624–2651, 2021.
- [2] Ram Zamir, Shlomo Shamai, and Uri Erez. Nested linear/lattice codes for structured multiterminal binning. *IEEE Transactions on Information Theory*, 48(6):1250–1276, 2002.

Weighted Parity-Check Codes (WPC)

In channel coding, the encoder encodes the message $\mathbf{m} \in \mathbb{F}_2^k$ into codeword $\mathbf{x} \in \mathbb{F}_2^n$. The decoder receives $\mathbf{y} \in \mathbb{F}_2^n$ and recovers $\hat{\mathbf{m}} \in \mathbb{F}_2^k$.

Randomly choose a full-rank parity check matrix $\mathbf{H} \in \mathbb{F}_2^{n \times n}$. For a bias vector $\mathbf{q} = [q_1, \dots, q_n] \in [0, 1]^n$, define the \mathbf{q} -weight of a vector $\mathbf{u} \in \mathbb{F}_2^n$ as

$$w_{\mathbf{q}}(\mathbf{u}) := \prod_{i=1}^{n} q_i^{u_i} (1 - q_i)^{1 - u_i} = 2^{-\sum_{i=1}^{n} H_b(u_i, q_i)}.$$

Intuitively, $w_{\mathbf{q}}(\mathbf{u})$ is the probability of \mathbf{u} assuming the entries $u_i \sim \text{Bern}(q_i)$ are independent across i.

Given the codeword/parity bias vectors $\mathbf{p}, \mathbf{q} \in [0, 1]^n$, the query function is

$$f_{\mathbf{H}}(\mathbf{p}, \mathbf{q}) := \operatorname{argmax}_{\mathbf{x} \in \mathbb{F}_2^n} w_{\mathbf{p}}(\mathbf{x}) w_{\mathbf{q}}(\mathbf{x} \mathbf{H}^T).$$
 (2)

The encoder has two parameters: the encoder codeword bias function \mathbf{p}_e : $\mathbb{F}_2^k \to [0,1]^n$ which maps the message $\mathbf{m} \in \mathbb{F}_2^k$ (and other information available at the encoder) to a bias vector $\mathbf{p}_e(\mathbf{m})$, and the encoder parity bias function $\mathbf{q}_e : \mathbb{F}_2^k \to [0,1]^n$. The actual encoding function is

$$\mathbf{m} \mapsto \mathbf{x} = f_{\mathbf{H}} \left(\mathbf{p}_e(\mathbf{m}), \, \mathbf{q}_e(\mathbf{m}) \right).$$

The decoder likewise has two parameters: the decoder codeword and parity bias functions $\mathbf{p}_d, \mathbf{q}_d : \mathbb{F}_2^n \to [0, 1]^n$. The decoding function is

$$\mathbf{y} \mapsto \hat{\mathbf{m}} = \left[(\hat{\mathbf{x}} \mathbf{H}^T)_1, \dots, (\hat{\mathbf{x}} \mathbf{H}^T)_k \right], \tag{2}$$

where $\hat{\mathbf{x}} := f_{\mathbf{H}}(\mathbf{p}_d(\mathbf{y}), \mathbf{q}_d(\mathbf{y})).$

Weighted Parity-Check Codes with State (WPCS)

The encoder observes \mathbf{m} , \mathbf{s} and uses $\mathbf{p}_e(\mathbf{m}, \mathbf{s})$, $\mathbf{q}_e(\mathbf{m}, \mathbf{s})$ to obtain \mathbf{x} . The decoder uses $\mathbf{p}_d(\mathbf{y})$, $\mathbf{q}_d(\mathbf{y})$ to obtain $\hat{\mathbf{x}}$, and outputs $\hat{\mathbf{m}} = [(\hat{\mathbf{x}}\mathbf{H}^T)_1, \dots, (\hat{\mathbf{x}}\mathbf{H}^T)_k]$.

$$\mathbf{p}_{e}(\mathbf{m}, \mathbf{s}) = [p_{e}(s_{1}), \dots, p_{e}(s_{n})],$$

$$\mathbf{q}_{e}(\mathbf{m}, \mathbf{s}) = [\mathbf{m}, \mathbf{q}],$$

$$\mathbf{p}_{d}(\mathbf{y}) = [p_{d}(y_{1}), \dots, p_{d}(y_{n})],$$

$$\mathbf{q}_{d}(\mathbf{y}) = [\frac{1}{2}\mathbf{1}^{k}, \mathbf{q}],$$

such that $\mathbf{q} = [q_1, \dots, q_{n-k}]$, where $q_i \sim P_Q$ i.i.d., and P_Q is a distribution over [0,1] symmetric about 1/2 (i.e., if $Q \sim P_Q$, then $1-Q \sim P_Q$).

Parity Bias Distribution

The nested linear code [2] is a special case of WPCS, where there are $n-k-\tilde{k}$ parity-check bits fixed to zero (i.e., $q_i=0$), and \tilde{k} unused parity-check bits $(q_i=1/2)$, where $\tilde{k}\in\{0,\ldots,n-k\}$ is the dimension of each coset. It can be approximated by taking $P_Q(0)=P_Q(1)=(1-\gamma)/2$, $P_Q(1/2)=\gamma$, where $\gamma=\tilde{k}/(n-k)$, giving around $(n-k)P_Q(1/2)=\tilde{k}$ unused parity-check bits.

We construct our P_Q so that (7) holds, named Threshold linear P_Q using a cdf:

$$F_{Q}(t) := \begin{cases} 0 & \text{if } t < 0 \\ \max\{\theta/2, 0\} & \text{if } 0 \le t < |\theta|/2 \\ t & \text{if } |\theta|/2 \le t < 1 - |\theta|/2 \\ 1 - \max\{\theta/2, 0\} & \text{if } 1 - |\theta|/2 \le t < 1 \\ 1 & \text{if } t \ge 1, \end{cases}$$
 (3)

where $\theta \in [-1, 1]$ is chosen such that (7) holds.

Optimality for the Channels with States

Consider WPCS that $|\mathcal{S}|, |\mathcal{Y}| < \infty$, and P_Q is a discrete distribution over [0,1] with finite support. Let $S \sim P_S$, $X|S \sim P_{X|S}$, $Y|(S,X) \sim P_{Y|S,X}$, $Q \sim P_Q$, $V \in \{0,1\}$, $V|Q \sim P_{V|Q}$, where $(P_{X|S}, P_{V|Q})$ is the minimizer of

$$\mathbf{E}[H_b(X, p_e(S))] + (1 - R)\mathbf{E}[H_b(V, Q)],$$
 (4)

where H_b is the binary cross entropy function, subject to

$$H(X|S) + (1-R)H(V|Q) \ge 1.$$
 (5)

If the minimizer of (4) is unique, and for all $P_{\tilde{X}|Y}$, $P_{\tilde{V}|Q}$ satisfying

$$H(\tilde{X}|Y) + (1-R)H(\tilde{V}|Q) \ge 1 - R,$$
 (6)

we have

 $\mathbf{E}[H_b(\tilde{X}, p_d(Y))] + (1-R)\mathbf{E}[H_b(\tilde{V}, Q)] > \mathbf{E}[H_b(X, p_d(Y))] + (1-R)\mathbf{E}[H_b(V, Q)]$ and then as $n \to \infty$, the probability of error tends to 0.

Corollary

Let $|\mathcal{S}|, |\mathcal{Y}| < \infty$, fix $P_{X|S}$. Consider WPCS that $p_e(s) = P_{X|S}(1|s)$, $p_d(y) = P_{X|Y}(1|y)$, and P_Q is discrete and over [0,1] with finite support satisfying

$$\mathbf{E}[H_b(Q)] = \frac{1 - H(X|S)}{1 - R}.\tag{7}$$

For any R < I(X;Y) - I(X;S), as $n \to \infty$, the error probability goes to 0.

Performance Evaluation

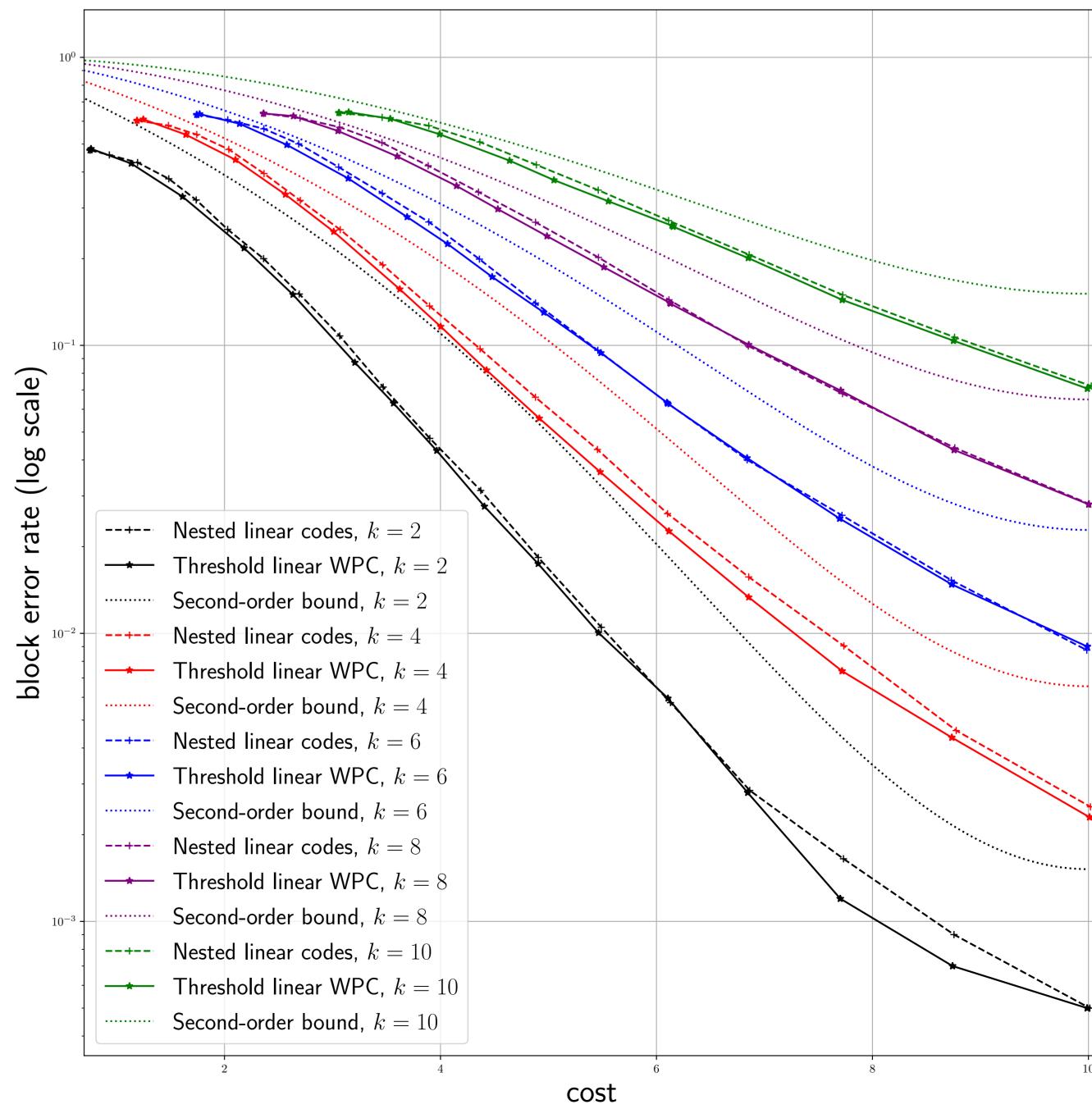


Figure 2. Performance of n=20, BSC channel of crossover probability $\beta=0.05$. Note we use $p_e(s)=\alpha^{1-s}(1-\alpha)^s$ for $S\to X$ be about BSC (α) , $p_d(y)=\beta^{1-y}(1-\beta)^y$ and (3) for P_Q .