**Weighted Parity-Check Codes** Chih Wei Ling **Yanxiao Liu** Cheuk Ting Li

Department of Information Engineering, The Chinese University of Hong Kong

 $\rightarrow \, \mathrm{m}$ 

## **Abstract**

We introduce a new class of codes, called weighted parity-check codes, where each parity-check bit has a weight that indicates its likelihood to be one (instead of fixing each parity-check bit to be zero). It is applicable to a wide range of settings, e.g. asymmetric channels, channels with state and/or cost constraints, and the Wyner-Ziv problem, and can provably achieve the capacity. For the channel with state (Gelfand-Pinsker) setting, our code not only achieves the capacity of any channel with state, but also achieves a smaller error rate compared to the nested linear code.

## **Ideas and Advantages**

The goal is to present a general code construction based on weighted code**book** idea, but with a linear structure for efficient encoding and decoding.

- The codebook is a "fuzzy set", where each bit sequence has a weight that corresponds to the likelihood that the sequence is selected.
- By [\[1\]](#page-0-0), weighted codebook eliminates the need of subcodebooks and gives sharper finite-blocklength and second-order error bounds.
- It applies to general(symmetric/asymmetric) channels with/without state.

Consider *s<sup>i</sup> iid* ∼ Bern(1*/*2) and *X* → *Y* is BSC(*β*). We have an expected cost/distortion constraint  $\mathbf{E}[|\{i : x_i \neq s_i\}|] \leq nD$ . For  $0 \leq \beta \leq D \leq 1/2$ , the capacity is the upper concave envelope of  $H(D) - H(\beta)$ .

## **Channels with State Information**

Consider the channel has a state sequence  $\mathbf{s} = [s_1, \ldots, s_n]$ , where  $s_i \in \mathcal{S}$ (not necessarily binary), *s<sup>i</sup> iid* ∼ *PS*, is available noncausally to the encoder. Given  $\mathbf{s}$ , the encoder encodes message  $\mathbf{m} \in \mathbb{F}_2^k$  $\frac{k}{2}$  into  $\mathbf{x} \in \mathbb{F}_2^n$  $\frac{n}{2}$ , which is sent through a memoryless channel  $P_{Y|S,X}(y|s,x)$ . The decoder receives y and outputs  $\hat{\mathbf{m}}$ . The input may have a cost constraint  $\mathbf{E}[\sum_{i=1}^{n}c(s_i,x_i)] \leq nD$ , where  $c : \mathcal{S} \times \mathbb{F}_2 \to [0, \infty)$ .

Intuitively,  $w_{q}(\mathbf{u})$  is the probability of **u** assuming the entries  $u_i \sim \text{Bern}(q_i)$ are independent across *i*.

Given the codeword/parity bias vectors  $\mathbf{p}, \mathbf{q} \in [0, 1]^n$ , the *query function* is  $f_{\mathbf{H}}(\mathbf{p}, \mathbf{q}) := \arg \max_{\mathbf{x} \in \mathbb{F}}$ *n*  $w_{\mathbf{p}}(\mathbf{x})w_{\mathbf{q}}(\mathbf{x}\mathbf{H}^T)$ )*.* (1)

The encoder has two parameters: the *encoder codeword bias function*  $\mathbf{p}_e$  :  $\mathbb{F}_2^k \ \rightarrow \ [0,1]^n$  which maps the message  $\mathbf{m} \ \in \ \mathbb{F}_2^k$  $_2^k$  (and other information available at the encoder) to a bias vector **p***e*(**m**), and the *encoder parity bias function*  $\mathbf{q}_e : \mathbb{F}_2^k \to [0,1]^n$ . The actual encoding function is

 $\mathbf{m} \mapsto \mathbf{x} = f_{\mathbf{H}}\left(\mathbf{p}_e(\mathbf{m}), \, \mathbf{q}_e(\mathbf{m})\right).$ 

## **Binary-Hamming Information Embedding**

The decoder likewise has two parameters: the *decoder codeword and parity bias functions*  $\mathbf{p}_d, \mathbf{q}_d : \mathbb{F}_2^n \to [0,1]^n$ . The decoding function is  $\mathbf{y} \mapsto \hat{\mathbf{m}} = \left[(\hat{\mathbf{x}} \mathbf{H}^T)_1, \dots, \ (\hat{\mathbf{x}} \mathbf{H}^T)_k\right]$ *,* (2) where  $\hat{\mathbf{x}} := f_{\mathbf{H}}\left(\mathbf{p}_d(\mathbf{y}), \, \mathbf{q}_d(\mathbf{y})\right)$ .



Figure 1. Block diagram of channel coding with state information. The encoder embeds m into the channel input **x**, which is with a cost constraint.  $\mathbf{e} \sim \text{Ber}(\beta)$  is the channel noise.

such that  $\mathbf{q} = [q_1, \ldots, q_{n-k}]$ , where  $q_i \sim P_Q$  i.i.d., and  $P_Q$  is a distribution over [0*,* 1] symmetric about 1*/*2 (i.e., if *Q* ∼ *PQ*, then 1 − *Q* ∼ *PQ*).

The nested linear code [\[2\]](#page-0-1) is a special case of WPCS, where there are *n*−*k*−˜*k* parity-check bits fixed to zero (i.e.,  $q_i = 0$ ), and  $\hat{k}$  unused parity-check bits  $\tilde{h}(q_i=1/2)$ , where  $\tilde{k}\in\{0,\ldots,n-k\}$  is the dimension of each coset. It can be approximated by taking  $P_Q(0) = P_Q(1) = (1 - \gamma)/2$ ,  $P_Q(1/2) = \gamma$ , where  $\gamma = \tilde{k}/(n-k)$ , giving around  $(n-k)P_Q(1/2) = \tilde{k}$  unused parity-check bits.

We construct our  $P_Q$  so that [\(7\)](#page-0-2) holds, named *Threshold linear*  $P_Q$  using a cdf:

### **References**

- <span id="page-0-0"></span>[1] Cheuk Ting Li and Venkat Anantharam. A unified framework for one-shot achievability via the poisson matching lemma. *IEEE Transactions on Information Theory*, 67(5):2624-2651, 2021.
- <span id="page-0-1"></span>[2] Ram Zamir, Shlomo Shamai, and Uri Erez. Nested linear/lattice codes for structured multiterminal binning. *IEEE Transactions on Information Theory*, 48(6):1250-1276, 2002.

 $t < 0$  $0 \le t < |\theta|/2$  $|\theta|/2 \le t < 1 - |\theta|/2$  $1 - |\theta|/2 \le t < 1$ if  $t \geq 1$ , (3)

## **Weighted Parity-Check Codes (WPC)**

In channel coding, the encoder encodes the message  $\mathbf{m} \in \mathbb{F}_2^k$ word  $\mathbf{x} \in \mathbb{F}_2^n$  $n_2^n$ . The decoder receives  $\mathbf{y} \in \mathbb{F}_2^n$ 

Randomly choose a full-rank parity check matrix  $\mathbf{H} \in \mathbb{F}_2^{n \times n}$ 2 . For a *bias*  $\mathbf{v}$  *ector*  $\mathbf{q} = [q_1, \ldots, q_n] \in [0, 1]^n$ , define the  $\mathbf{q}$ *-weight* of a vector  $\mathbf{u} \in \mathbb{F}_2^n$  $\frac{n}{2}$  as

Consider WPCS that  $|S|$ ,  $[0, 1]$  with finite support. *Q* ∼ *P*<sub>*Q*</sub>, *V* ∈ {0, 1}, *V* |*Q*  $\mathbf{E}$  [ $H_b(X,$ If the minimizer of [\(4\)](#page-0-3) is unique  $H(X|Y)$ we have

Let  $|S|, |Y| < \infty$ , fix  $P_{X|S}$ . Consider WPCS that  $p_e(s) = P_{X|S}(1|s)$ ,  $p_d(y) = 0$  $P_{X|Y}(1|y)$ , and  $P_Q$  is discrete and over [0, 1] with finite support satisfying  $\mathbf{E}[H_b(Q)] = % \begin{cases} \sum_{l=0}^m\frac{1}{l}\sum_{l=0}^{L-1} \left\vert \mathcal{A}_l(Q_l)\right\vert \leq 1, \ \sum_{l=0}^m\frac{1}{l}\sum_{l=0}^{L-1} \left\vert \mathcal{A}_l(Q_l)\right\vert \leq 1, \ \sum_{l=0}^m\frac{1}{l}\sum_{l=0}^{L-1} \left\vert \mathcal{A}_l(Q_l)\right\vert \leq 1, \ \sum_{l=0}^m\frac{1}{l}\sum_{l=0}^{L-1} \left\vert \mathcal{A}_l(Q_l)\right\vert \leq 1,$  $1 - H(X|S)$  $1 - R$ *.* (7) For any  $R < I(X;Y) - I(X;S)$ , as  $n \to \infty$ , the error probability goes to 0.



$$
w_{\mathbf{q}}(\mathbf{u}) := \prod_{i=1}^{n} q_i^{u_i} (1 - q_i)^{1 - u_i}
$$

# **Weighted Parity-Check Codes with State (WPCS)**

The encoder observes  $\mathbf{m}$ ,  $\mathbf{s}$  and uses  $\mathbf{p}_e(\mathbf{m}, \mathbf{s})$ ,  $\mathbf{q}_e(\mathbf{m}, \mathbf{s})$  to obtain  $\mathbf{x}$ . The decoder uses  $\mathbf{p}_d(\mathbf{y}), \mathbf{q}_d(\mathbf{y})$  to obtain  $\hat{\mathbf{x}},$  and outputs  $\hat{\mathbf{m}} = [(\hat{\mathbf{x}} \mathbf{H}^T)_1, \dots,$   $(\hat{\mathbf{x}} \mathbf{H}^T)_k].$  $p_e(s_n)],$ 

$$
\mathbf{p}_e(\mathbf{m}, \mathbf{s}) = [p_e(s_1), \ldots
$$

$$
\mathbf{q}_e(\mathbf{m}, \mathbf{s}) = [\mathbf{m}, \, \mathbf{q}],
$$

 $\mathbf{p}_d(\mathbf{y}) = [p_d(y_1), \dots, p_d(y_n)],$ 

$$
\mathbf{q}_d(\mathbf{y}) = [\frac{1}{2}\mathbf{1}^k, \, \mathbf{q}],
$$

# **Parity Bias Distribution**

$$
F_Q(t) := \begin{cases} 0 & \text{if} \\ \max\{\theta/2, 0\} & \text{if} \\ 1 - \max\{\theta/2, 0\} & \text{if} \\ 1 & \text{if} \end{cases}
$$

where  $\theta \in [-1, 1]$  is chosen such that [\(7\)](#page-0-2) holds.

 $_2^k$  into code- $\hat{\textbf{z}}_2^n$  and recovers  $\hat{\textbf{m}} \in \mathbb{F}_2^k$  $\frac{k}{2}$ .

 $1-u_i = 2^{-\sum_{i=1}^n H_b(u_i,q_i)}$ .

## **Optimality for the Channels with States**



 $\mathbf{E}[H_b(\tilde{X},p_d(Y))] + (1-R)\mathbf{E}[H_b(\tilde{V},Q)] > \mathbf{E}[H_b(X,p_d(Y))] + (1-R)\mathbf{E}[H_b(V,Q)]$ and then as  $n \to \infty$ , the probability of error tends to 0.

## <span id="page-0-3"></span><span id="page-0-2"></span>**Corollary**

### **Performance Evaluation**

<span id="page-0-4"></span>Figure 2. Performance of  $n = 20$ , BSC channel of crossover probability  $\beta = 0.05$ . Note we use  $p_e(s)=\alpha^{1-s}(1-\alpha)^s$  for  $S\to X$  be about BSC( $\alpha$ ),  $p_d(y)=\beta^{1-y}(1-\beta)^y$  and [\(3\)](#page-0-4) for  $P_Q.$ 

